# **ASSP**

# Dual Serial Input PLL Frequency Synthesizer

# MB15F02SL

#### n **DESCRIPTION**

The Fujitsu MB15F02SL is a serial input Phase Locked Loop (PLL) frequency synthesizer with a 1200 MHz and a 500 MHz prescalers. The 1200 MHz and 500 MHz prescalers have a dual modulus division ratio of 128/129 or 64/65, and a 8/9 or a 16/17 enabling pulse swallowing operation.

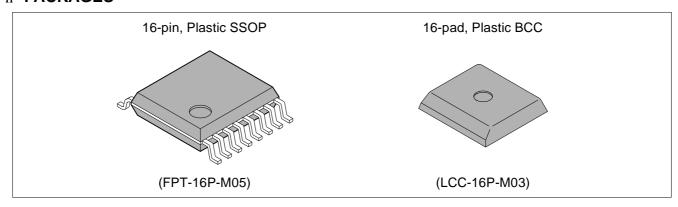
The supply voltage range is between 2.4 V and 3.6 V. The MB15F02SL uses the latest BiCMOS process. As a result, the supply current is typically 3 mA at 2.7 V. A refined charge pump supplies a well-balanced output current of 1.5 mA or 6 mA. The charge pump current is selectable by serial data.

MB15F02SL is ideally suited for wireless mobile communications, such as GSM and PDC.

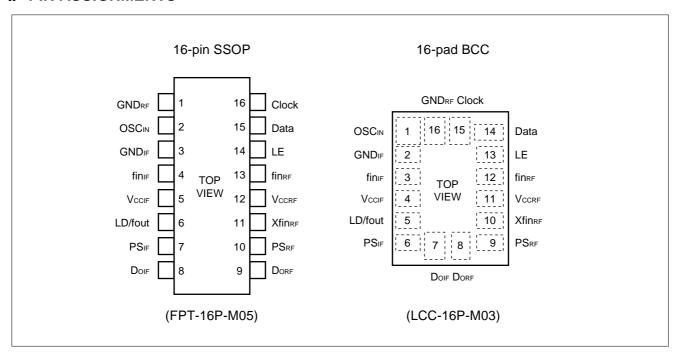
#### n FEATURES

- High frequency operation: RF synthesizer: 1200 MHz max
   IF synthesizer: 500 MHz max
- Low power supply voltage: Vcc = 2.4 to 3.6 V
- Ultra Low power supply current: Icc = 3.0 mA typ. (Vcc = 2.7 V, Ta = +25°C, in IF, RF locking state) Icc = 3.5 mA typ. (Vcc = 3.0 V, Ta = +25°C, in IF, RF locking state)
- Direct power saving function: Power supply current in power saving mode
  - Typ.  $0.1 \,\mu\text{Å} \,(\text{Vcc} = \text{Vp} = 3.0 \,\text{V}, \, \text{Ta} = +25^{\circ}\text{C}), \, \text{Max.} \, 10 \,\mu\text{A} \,(\text{Vcc} = \text{Vp} = 3.0 \,\text{V})$
- Dual modulus prescaler: 1200 MHz prescaler (64/65, 128/129)/500 MHz prescaler (8/9 or 16/17)
- Serial input 14-bit programmable reference divider: R = 3 to 16,383
- Serial input programmable divider consisting of:
  - Binary 7-bit swallow counter: 0 to 127
  - Binary 11-bit programmable counter: 3 to 2,047
- Software selectable charge pump current
- · On-chip phase control for phase comparator
- Operating temperature: Ta = -40 to +85°C
- Pin compatible with MB15F02, MB15F02L

#### n PACKAGES



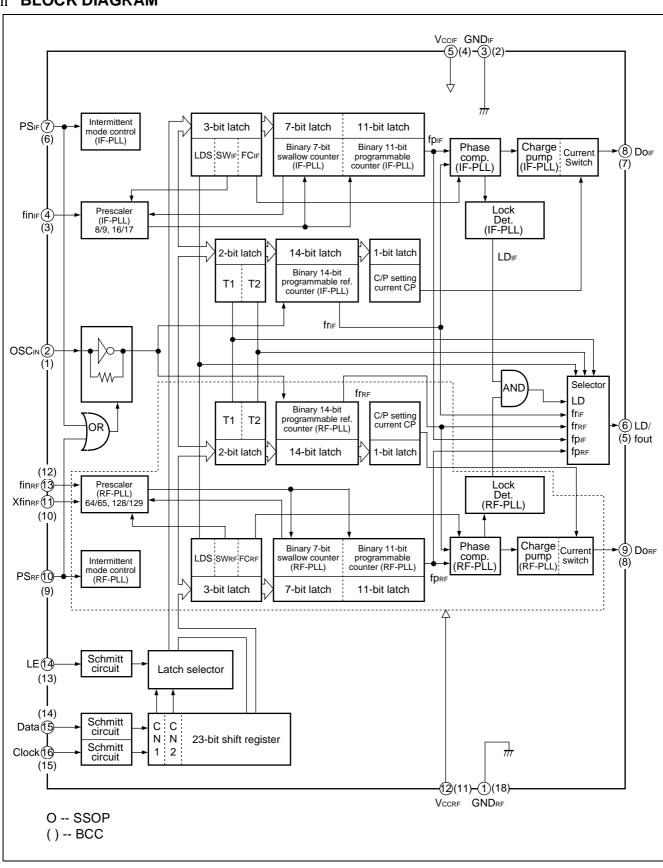
#### n PIN ASSIGNMENTS



#### n PIN DESCRIPTIONS

Pin	no.	Pin	1/0	D
SSOP	всс	name	I/O	Descriptions
1	16	GNDRF	_	Ground for RF-PLL section.
2	1	OSCIN	I	The programmable reference divider input. TCXO should be connected with a AC coupling capacitor.
3	2	GND <sub>IF</sub>	_	Ground for the IF-PLL section.
4	3	finıF	I	Prescaler input pin for the IF-PLL. Connection to an external VCO should be via AC coupling.
5	4	Vccif	_	Power supply voltage input pin for the IF-PLL section.
6	5	LD/fout	0	Lock detect signal output (LD)/phase comparator monitoring output (fout). The output signal is selected by LDS bit in a serial data. LDS bit = "H"; outputs fout signal LDS bit = "L"; outputs LD signal
7	6	PSıF	I	Power saving mode control for the IF-PLL section. This pin must be set at "L" during Power-ON. (Open is prohibited.) PS <sub>IF</sub> = "H"; Normal mode PS <sub>IF</sub> = "L"; Power saving mode
8	7	Doif	0	Charge pump output for the IF-PLL section. Phase characteristics of the phase detector can be selected via programming of the FC-bit.
9	8	Dorf	0	Charge pump output for the RF-PLL section. Phase characteristics of the phase detector can be selected via programming of the FC-bit.
10	9	PSRF	I	Power saving mode control for the RF-PLL section. This pin must be set at "L" during Power-ON. (Open is prohibited.) PSRF = "H"; Normal mode PSRF = "L"; Power saving mode
11	10	Xfinrf	I	Prescaler complementary input for the RF-PLL section. This pin should be grounded via a capacitor.
12	11	Vccrf	_	Power supply voltage input pin for the RF-PLL section, the shift register and the oscillator input buffer. When power is OFF, latched data of RF-PLL is lost.
13	12	fin <sub>RF</sub>	I	Prescaler input pin for the RF-PLL. Connection to an external VCO should be via AC coupling.
14	13	LE	I	Load enable signal inpunt (with a schmitt trigger input buffer.) When the LE bit is set "H", data in the shift register is transferred to the corresponding latch according to the control bit in the serial data.
15	14	Data	I	Serial data input (with a schmitt trigger input buffer.) Data is transferred to the corresponding latch (IF-ref counter, IF-prog. counter, RF-ref. counter, RF-prog. counter) according to the control bit in the serial data.
16	15	Clock	I	Clock input for the 23-bit shift register (with a schmitt trigger input buffer.) One bit of data is shifted into the shift register on a rising edge of the clock.

#### n BLOCK DIAGRAM



#### n ABSOLUTE MAXIMUM RATINGS

Parameter	Symbol	Ra	ting	Unit	Remark
Faranielei	Symbol	Min.	Max.	Oille	Remark
Power supply voltage	Vcc	-0.5	+4.0	V	
Input voltage	Vı	-0.5	Vcc +0.5	V	
Output voltage	Vo	GND	Vcc	V	
Storage temperature	Tstg	-55	+125	°C	

WARNING: Semiconductor devices can be permanently damaged by application of stress (voltage, current, temperature, etc.) in excess of absolute maximum ratings. Do not exceed these ratings.

#### n RECOMMENDED OPERATING CONDITIONS

Parameter	Sumbol		Value	Unit	Remark	
Parameter	Symbol	Min.	Тур.	Max.	Unit	Remark
Power supply voltage	Vcc	2.4	3.0	3.6	V	
Input voltage	Vı	GND	_	Vcc	V	
Operating temperature	Та	-40	_	+85	°C	

WARNING: The recommended operating conditions are required in order to ensure the normal operation of the semiconductor device. All of the device's electrical characteristics are warranted when the device is operated within these ranges.

Always use semiconductor devices within their recommended operating condition ranges. Operation outside these ranges may adversely affect reliability and could result in device failure.

No warranty is made with respect to uses, operating conditions, or combinations not represented on the data sheet. Users considering application outside the listed conditions are advised to contact their FUJITSU representatives beforehand.

#### n ELECTRICAL CHARACTERISTICS

 $(Vcc = 2.4 \text{ V to } 3.6 \text{ V}, Ta = -40 \text{ to } +85^{\circ}\text{C})$ 

Davamatar		Comple of	Condition		Value		1110:4
Parameter		Symbol	Condition	Min.	Тур.	Max.	Unit
Power supply current*1		CCIF*1	finif = 500 MHz, Vccif = 2.7 V (Vccif = 3.0 V)	_	1.2 (1.5)	_	mA
rower supply current		CCRF*1	finrf = 1200 MHz, Vccrf = 2.7 V (Vccrf = 3.0 V)	_	1.8 (2.0)	_	mA
Power saving current		<b>I</b> PSIF	PS <sub>IF</sub> = PS <sub>RF</sub> = "L"	_	0.1*2	10	μΑ
Fower saving current		IPSRF	PS <sub>IF</sub> = PS <sub>RF</sub> = "L"	_	0.1*2	10	μА
	fin⊮*³	finıғ	IF PLL	50	_	500	MHz
Operating frequency	fin <sub>RF</sub> *3	finre	RF PLL	100	_	1200	MHz
	OSCIN	fosc	_	3	_	40	MHz
	fin⊮*8	Vfinif	IF PLL, 50 Ω system	-15	_	+2	dBm
Input sensitivity	finrf	Vfinrf	RF PLL, 50 Ω system	-15	_	+2	dBm
	OSCIN	Vosc	_	0.5		Vcc	Vp-p
"H" level input voltage	Data, Clock,	Vıн	Schmitt trigger input	Vcc × 0.7 + 0.4	_	-	V
"L" level input voltage	LE	VıL	Schmitt trigger input	_	_	Vcc × 0.3 - 0.4	V
"H" level input voltage	PS	Vін	_	Vcc × 0.7	_	_	V
"L" level input voltage	FS	VIL	_	_	_	Vcc×0.3	\ \
"H" level input current	Data,	Iıн*⁴	_	-1.0	_	+1.0	
"L" level input current	Clock, LE, PS	IIL*4	-	-1.0	_	+1.0	μΑ
"H" level input current	OSCIN	Іін	-	0	_	+100	
"L" level input current	OSCIN	Iı∟*4	_	-100	_	0	μΑ
"H" level output volt- age	LD/fout	Vон	Vcc = 3.0 V, Iон = -1 mA	Vcc - 0.4	_	_	V
"L" level output volt- age	LD/IOut	Vol	Vcc = 3.0 V, loL = 1 mA	_	_	0.4	v
"H" level output volt- age	Doif	V <sub>DOH</sub>	Vcc = 3.0 V, IDOH = -0.5 mA	Vcc - 0.4	-	_	V
"L" level output volt- age	Dorf	Vdol	Vcc = 3.0 V, IDOL = 0.5 mA	_	_	0.4	V
High impedance cutoff current	Doif Dorf	loff	Vcc = 3.0 V, Voff = 0.5 V to Vcc - 0.5 V	_	_	2.5	nA
"H" level output cur- rent	LD/fout	Іон*4	Vcc = 3.0 V	-1.0	_	_	mA
"L" level output cur- rent	LD/IOUt	lo <sub>L</sub> *4	Vcc = 3.0 V	_	_	1.0	

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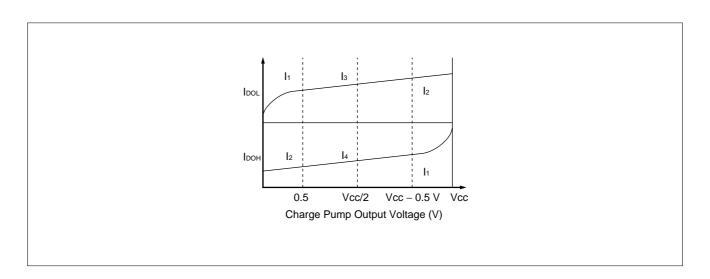
 $(Vcc = 2.4 \text{ to } 3.6 \text{ V}, Ta = -40 \text{ to } +85^{\circ}\text{C})$ 

Parameter		Symbol	Con	dition		Value		Unit
Farameter		Symbol	Com	uition	Min.	Тур.	Max.	Ullit
"I I" lavel autout avenue		. *4	Vcc = 3.0  V,	CS bit = "H"	_	-6.0	_	
"H" level output current	Doif	IDOH*4	$V_{DOH} = V_{CC}/2,$ $Ta = +25^{\circ}C$	CS bit = "L"	_	-1.5	_	mA
(1.2.11	DORF ent	IDOL	Vcc = 3.0 V,	CS bit = "H"	_	6.0	_	IIIA
"L" level output current			V <sub>DOL</sub> = V <sub>CC</sub> /2, Ta = +25°C	CS bit = "L"	_	1.5	_	
	I <sub>DOL</sub> /I <sub>DOH</sub>	I <sub>DOMT</sub> *5	$V_{DO} = V_{CC}/2$	•	_	3	_	%
Charge pump	vs V <sub>DO</sub>	IDOVD*6	0.5 V ≤ V <sub>DO</sub> ≤	Vcc – 0.5 V	_	10	_	%
current rate	vs Ta	IDOTA*7	-40°C ≤ Ta ≤ VDO = Vcc/2	+85°C,	_	10	_	%

- \*1: Conditions; fosc = 12 MHz, Ta = +25°C, in locking state.
- \*2:  $V_{CCIF} = V_{CCRF} = 3.0 \text{ V}$ , fosc = 12.8 MHz, Ta = +25°C, in power saving mode.
- \*3: AC coupling. 1000pF capacitor is connected under the condition of min. operating frequency.
- \*4: The symbol "-" (minus) means direction of current flow.
- \*5: Vcc = 3.0 V,  $Ta = +25^{\circ}\text{C} (||3| ||4|)/[(||3| + ||4|)/2] \times 100(\%)$ \*6: Vcc = 3.0 V,  $Ta = +25^{\circ}\text{C} [(||2| ||1|)/2]/[(||1| + ||2|)/2] \times 100(\%)$  (Applied to each IDOL, IDOH)
- \*7: Vcc = 3.0 V, [|Ido(85°C) Ido(-40°C)|/2]/[|Ido(85°C) + Ido(-40°C)|/2] × 100(%) (Applied to each Idol, Idoh)

	100 0.0 1, [[1.50(00 0)	150( 100)[, = ], [[150(000), 150(1000)]		···,
*8:	Prescaler divided ratio	Charge pump current	finı⊧	Vfin⊮(min)
	16/17	1.5 mA mode	$50 \text{ MHz} \le \text{fin} \le 500 \text{ MHz}$	–15 dBm
		6.0 mA mode	$50 \text{ MHz} \le \text{fin} \le 300 \text{ MHz}$	–15 dBm
			300 MHz < fin $\leq$ 500 MHz	–10 dBm
	8/9	1.5 mA mode	50 MHz $\leq$ fin $\leq$ 300 MHz*	–15 dBm
			300 MHz < fin $\leq$ 500 MHz	–15 dBm
		6.0 mA mode	50 MHz $\leq$ fin $\leq$ 300 MHz*	–15 dBm
			$300 \text{ MHz} < \text{fin} \le 500 \text{ MHz}$	_10 dBm

\*: Vcc = 2.7 V to 3.6 V at 500 MHz, $Vcc = 2.4 \text{ V to } 3.6 \text{ V}, \text{ Ta} = -40^{\circ}\text{C to } +85^{\circ}\text{C at fin} < 500 \text{ MHz}$ 



#### n FUNCTIONAL DESCRIPTION

The divide ratio can be calculated using the following equation:

 $f_{VCO} = \{(M \times N) + A\} \times f_{OSC} \div R \quad (A < N)$ 

fvco: Output frequency of external voltage controlled oscillator (VCO)

M : Preset divide ratio of dual modulus prescaler (8 or 16 for IF-PLL, 64 or 128 for RF-PLL)

N : Preset divide ratio of binary 11-bit programmable counter (3 to 2,047)

A : Preset divide ratio of binary 7-bit swallow counter ( $0 \le A \le 127$ )

fosc: Reference oscillation frequency

R : Preset divide ratio of binary 14-bit programmable reference counter (3 to 16,383)

#### **Serial Data Input**

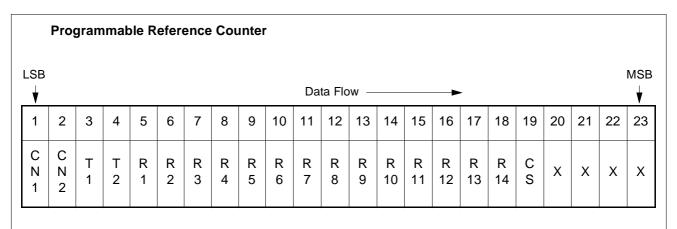
Serial data is entered using three pins, Data pin, Clock pin, and LE pin. Programmable dividers of IF/RF-PLL sections, programmable reference dividers of IF/RF-PLL sections are controlled individually. Serial data of binary data is entered through Data pin.

On rising edge of Clock, one bit of serial data is transferred into the shift register. When the LE signal is taken high, the data stored in the shift register is transferred to one of latch of them depending upon the control bit data setting.

**Table 1. Control Bit** 

Con	trol bit	Destination of serial data					
CN1	CN2	Destination of Serial data					
L	L	The programmable reference counter for the IF-PLL					
Н	L	The programmable reference counter for the RF-PLL					
L	Н	The programmable counter and the swallow counter for the IF-PLL					
Н	Н	The programmable counter and the swallow counter for the RF-PLL					

#### **Shift Register Configuration**



CN1,2 : Control bit [Table 1]

R1 to R14 : Divide ratio setting bits for the programmable reference counter (5 to 16,383)[Table 2]
T1, 2 : Test purpose bit [Table 3]
CS : Charge pump currnet select bit [Table 9]

X : Dummy bits (Set "0" or "1")

NOTE: Data input with MSB first.

	Pro	gran	nmal	ole C	oun	ter																
SB										Dat	a Flov	w —	-									MSE
1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19	20	21	22	23
C N 1	C N 2	L D S	S W IF/RF	F C IF/RF	A 1	A 2	A 3	A 4	A 5	A 6	A 7	N 1	N 2	N 3	N 4	N 5	N 6	N 7	N 8	N 9	N 10	N 11
$oxed{  N   N   D   W   C   1   2   3   4   5   6   7   1   2   3   4   5   6   7   8   9   10  }$																						

Table 2. Binary 14-bit Programmable Reference Counter Data Setting

Divide ratio (R)	R 14	R 13	R 12	R 11	R 10	R 9	R 8	R 7	R 6	R 5	R 4	R 3	R 2	R 1
3	0	0	0	0	0	0	0	0	0	0	0	0	1	1
4	0	0	0	0	0	0	0	0	0	0	0	1	0	0
			·	·			•	•	•	•	·		•	
16383	1	1	1	1	1	1	1	1	1	1	1	1	1	1

Note: • Divide ratio less than 3 is prohibited.

**Table 3. Test Purpose Bit Setting** 

T 1	T 2	LD/fout pin state
L	L	Outputs frif.
Н	L	Outputs fr <sub>RF</sub> .
L	Н	Outputs fp⊩.
Н	Н	Outputs fprf.

Table 4. Binary 11-bit Programmable Counter Data Setting

Divide ratio (N)	N 11	N 10	N 9	N 8	N 7	N 6	N 5	N 4	N 3	N 2	N 1
3	0	0	0	0	0	0	0	0	0	1	1
4	0	0	0	0	0	0	0	0	1	0	0
				•	•	•		•	•		•
2047	1	1	1	1	1	1	1	1	1	1	1

Note: • Divide ratio less than 3 is prohibited.

Table 5. Binary 7-bit Swallow Counter Data Setting

Divide ratio (N)	A 7	A 6	A 5	A 4	A 3	A 2	A 1
0	0	0	0	0	0	0	0
1	0	0	0	0	0	0	1
	,	•		•	,	•	
127	1	1	1	1	1	1	1

Note: • Divide ratio (A) range = 0 to 127

**Table 6. Prescaler Data Setting** 

		SW = "H"	SW = "L"
Prescaler divide ratio	IF-PLL	8/9	16/17
	RF-PLL	64/65	128/129

**Table 7. Phase Comparator Phase Switching Data Setting** 

	FCIF, RF = "H"	FCif, RF = "L"	
	DOIF, RF		
fr > fp	Н	L	
fr = fp	Z	Z	
fr < fp	L	Н	
VCO polarity	(1)	(2)	

VCO Output Frequency

LPF Output Voltage 

(1)

Note: • Z = High-impedance

• Depending upon the VCO and LPF polarity, FC bit should be set.

Table 8. LD/fout Output Select Data Setting

LDS	LD/fout output signal	
Н	fout (frif/RF, fpif/RF) signals	
L	LD signal	

#### **Table 9. Charge Pump Current Setting**

CS	Current value	
Н	±6.0 mA	
L	±1.5 mA	

#### **Power Saving Mode (Intermittent Mode Control Circuit)**

Table 10. PS Pin Setting

PS pin	Status	
Н	Normal mode	
L	Power saving mode	

The intermittent mode control circuit reduces the PLL power consumption.

By setting the PS pin low, the device enters into the power saving mode, reducing the current consumption. See the Electrical Characteristics chart for the specific value.

The phase detector output, Do, becomes high impedance.

For the dual PLL, the lock detector, LD, is as shown in the LD Output Logic table.

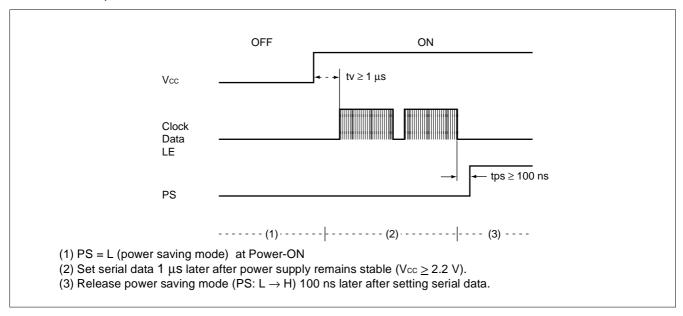
Setting the PS pin high, releases the power saving mode, and the device works normally.

The intermittent mode control circuit also ensures a smooth startup when the device returns to normal operation. When the PLL is returned to normal operation, the phase comparator output signal is unpredictable. This is because of the unknown relationship between the comparison frequency (fp) and the reference frequency (fr) which can cause a major change in the comparator output, resulting in a VCO frequency jump and an increase in lockup time.

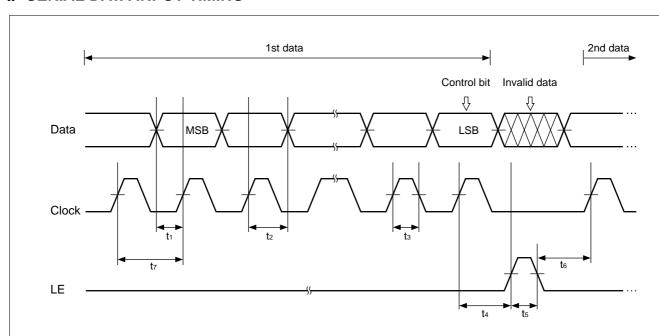
To prevent a major VCO frequency jump, the intermittent mode control circuit limits the magnitude of the error signal from the phase detector when it returns to normal operation.

Note: • When power (Vcc) is first applied, the device must be in standby mode, PS = Low, for at least 1 μs.

Note: • PS pin must be set at "L" for Power-ON.



#### n SERIAL DATA INPUT TIMING



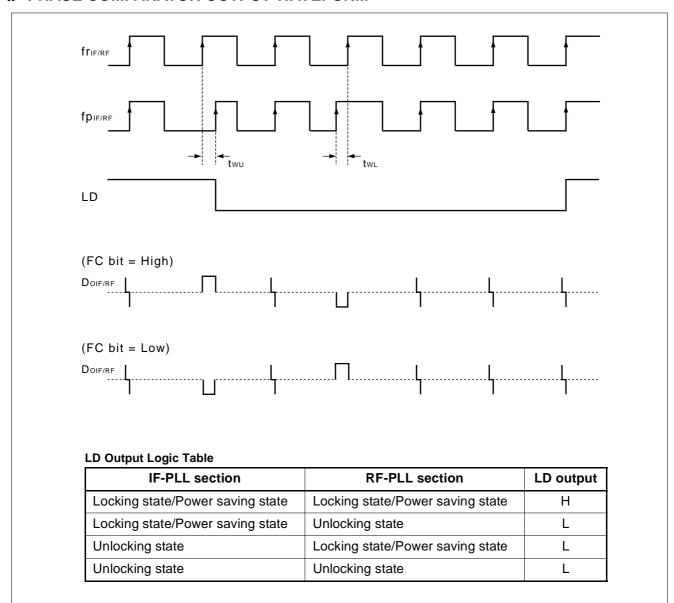
On rising edge of the clock, one bit of the data is transfered into the shift register.

Parameter	Min.	Тур.	Max.	Unit
t <sub>1</sub>	20	_	_	ns
t <sub>2</sub>	20	_	_	ns
t <sub>3</sub>	30	_	_	ns
t <sub>4</sub>	30	_	_	ns

Parameter	Min.	Тур.	Max.	Unit
<b>t</b> 5	100	_	_	ns
<b>t</b> 6	20	_	_	ns
<b>t</b> 7	100	_	_	ns

Note: LE should be "L" when the data is transferred into the shift register.

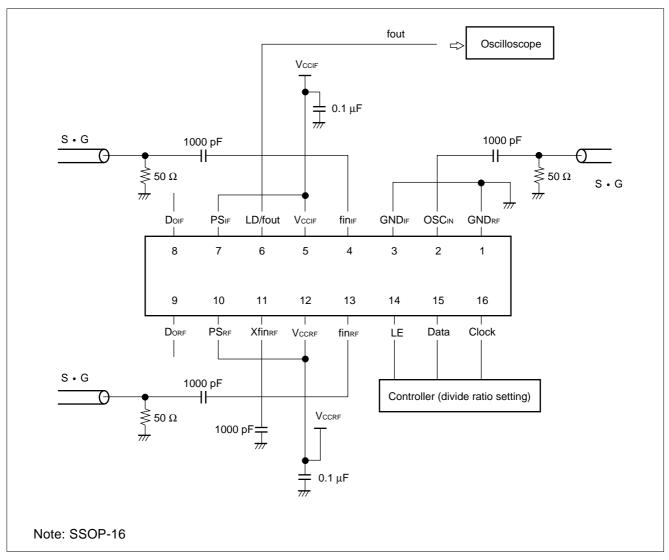
#### n PHASE COMPARATOR OUTPUT WAVEFORM



Notes: • Phase error detection range =  $-2\pi$  to  $+2\pi$ 

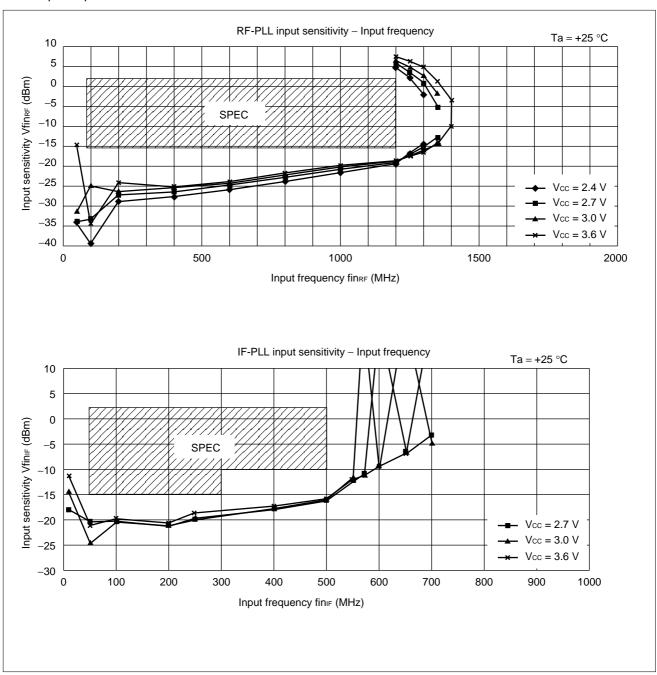
- Pulses on DoiF/RF signals are output to prevent dead zone.
- LD output becomes low when phase error is two or more.
- LD output becomes high when phase error is twL or less and continues to be so for three cycles or more.
- twu and twL depend on OSC<sub>IN</sub> input frequency as follows.
   twu ≥ 2/fosc: i. e. twu ≥ 156.3 ns when foscin = 12.8 MHz
   twu ≤ 4/fosc: i. e. twL ≤ 312.5 ns when foscin = 12.8 MHz

#### n MEASURMENT CIRCUIT (for Measuring Input Sensitivity fin/OSCin)

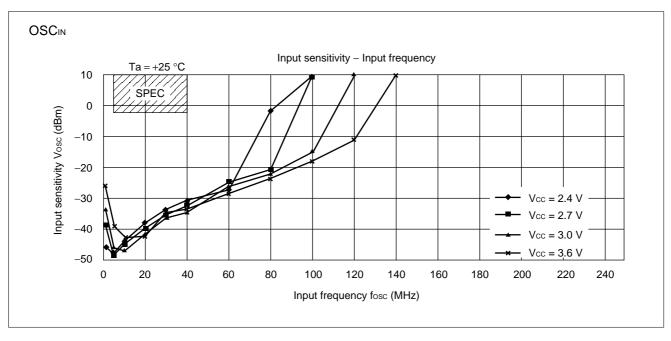


#### **n TYPICAL CHARACTERISTICS**

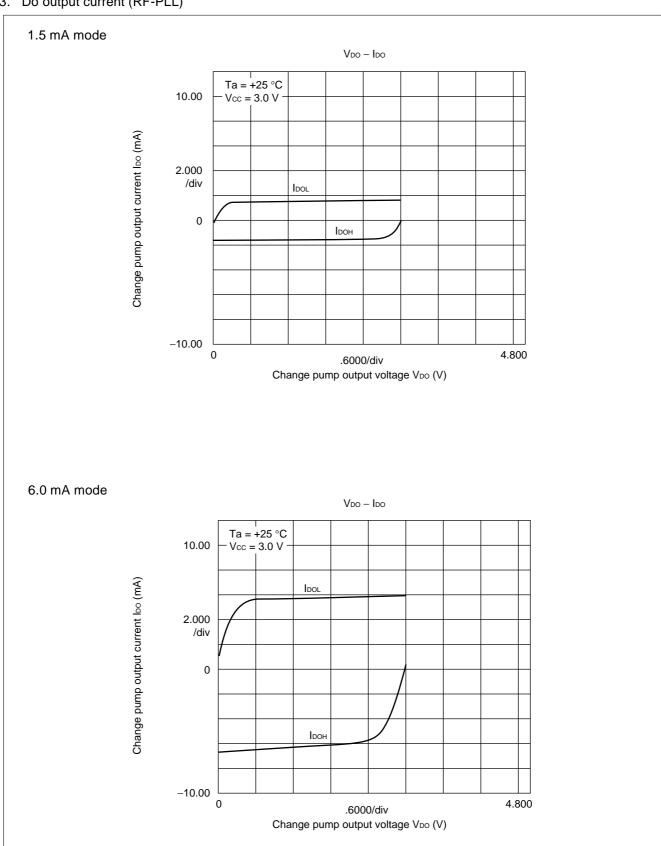
#### 1. fin input impedance

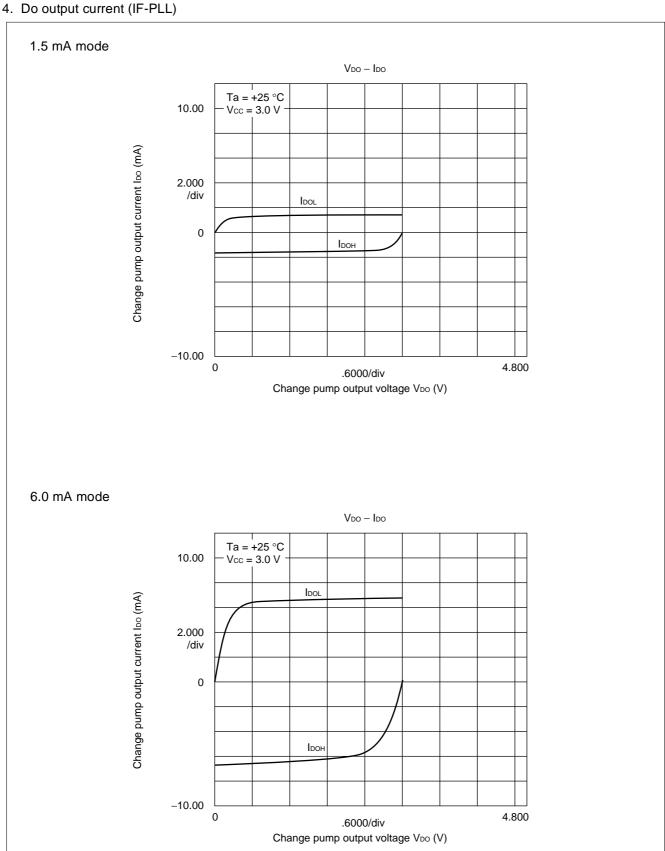


#### 2. OSC<sub>IN</sub> input sensitivity

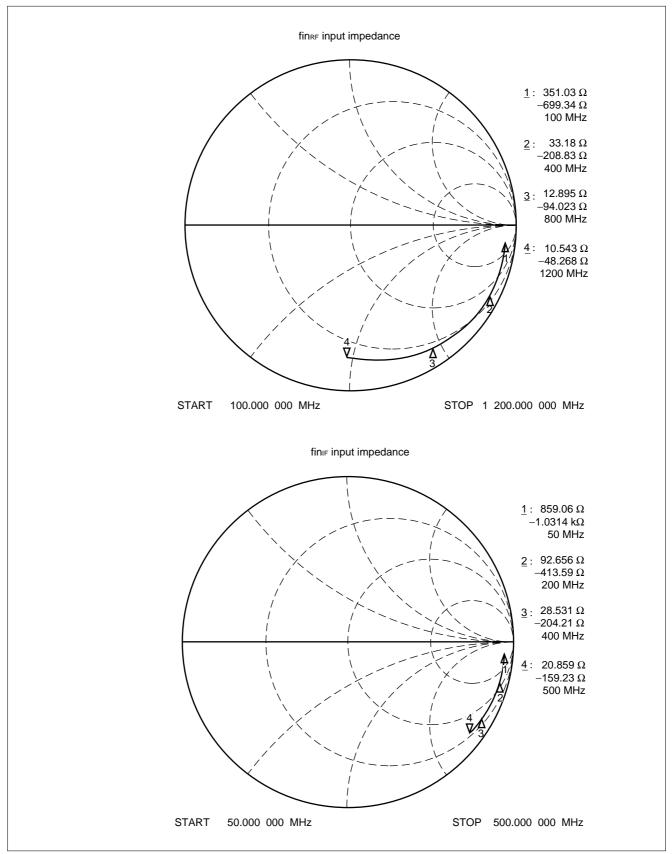


#### 3. Do output current (RF-PLL)

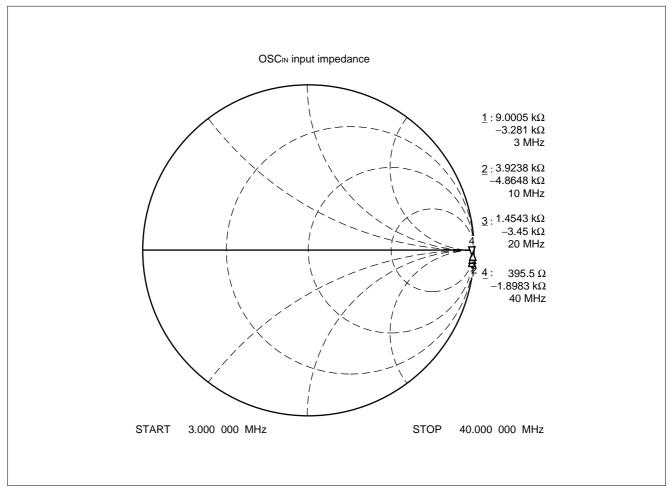




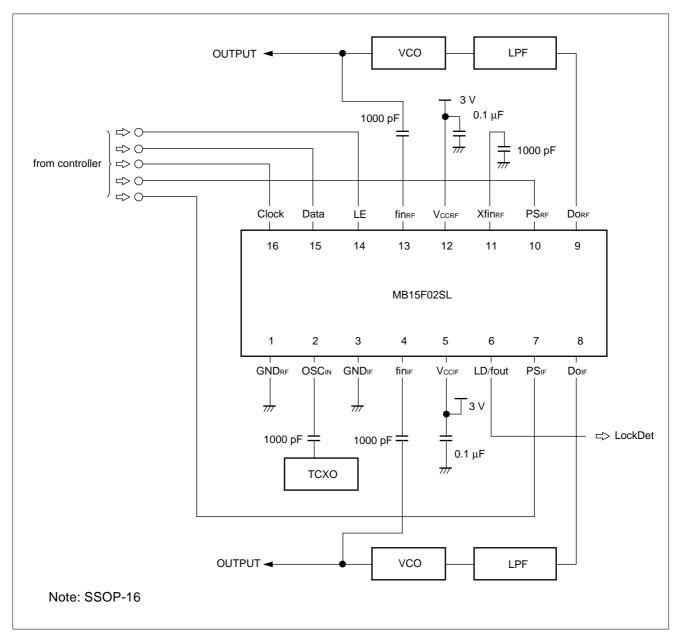
#### 5. fin input impedance



#### 6. OSC<sub>IN</sub> input impedance



#### n APPLICATION EXAMPLE



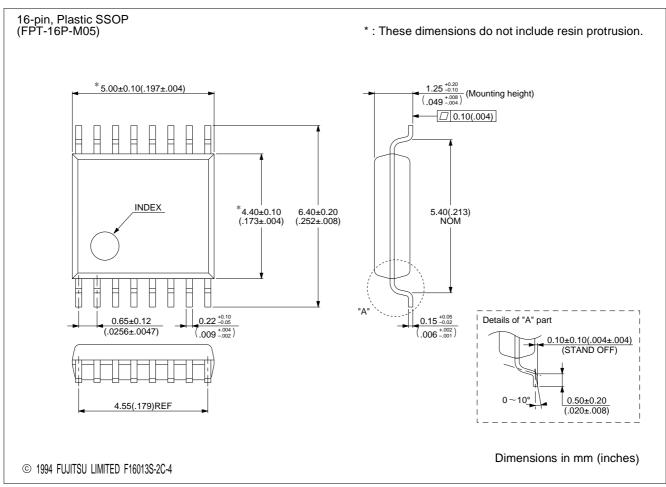
#### n USAGE PRECAUTIONS

- (1) VCCRF must equal VccIF.
  - Even if either RF-PLL or IF-PLL is not used, power must be supplied to both VccrF and VccrF to keep them equal. It is recommended that the non-use PLL is controlled by power saving function.
- (2) To protect against damage by electrostatic discharge, note the following handling precautions:
  - -Store and transport devices in conductive containers.
  - -Use properly grounded workstations, tools, and equipment.
  - -Turn off power before inserting or removing this device into or from a socket.
  - -Protect leads with conductive sheet, when transporting a board mounted device.

#### n ORDERING INFORMATION

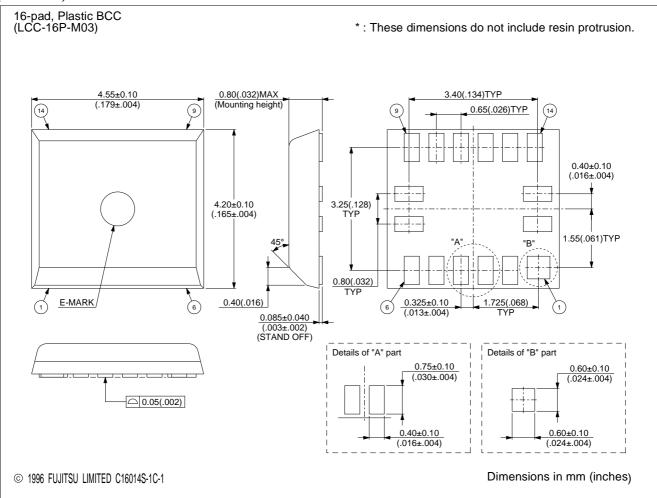
Part number	Package	Remarks
MB15F02SLPFV1	16-pin, plastic SSOP (FPT-16P-M05)	
MB15F02SLPV	16-pad, plastic BCC (LCC-16P-M03)	

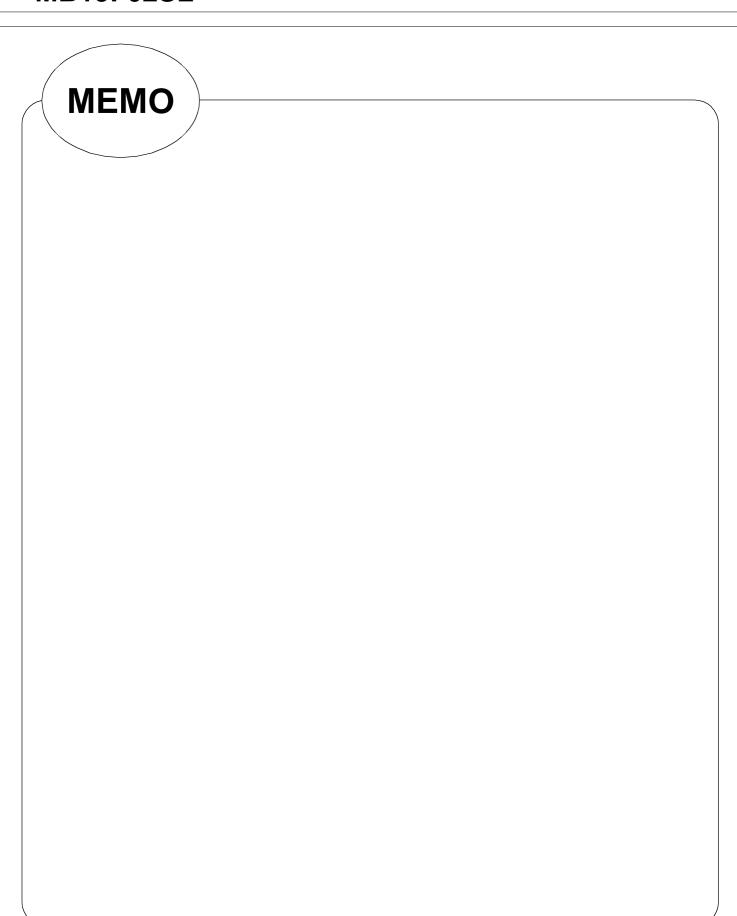
#### n PACKAGE DIMENSIONS

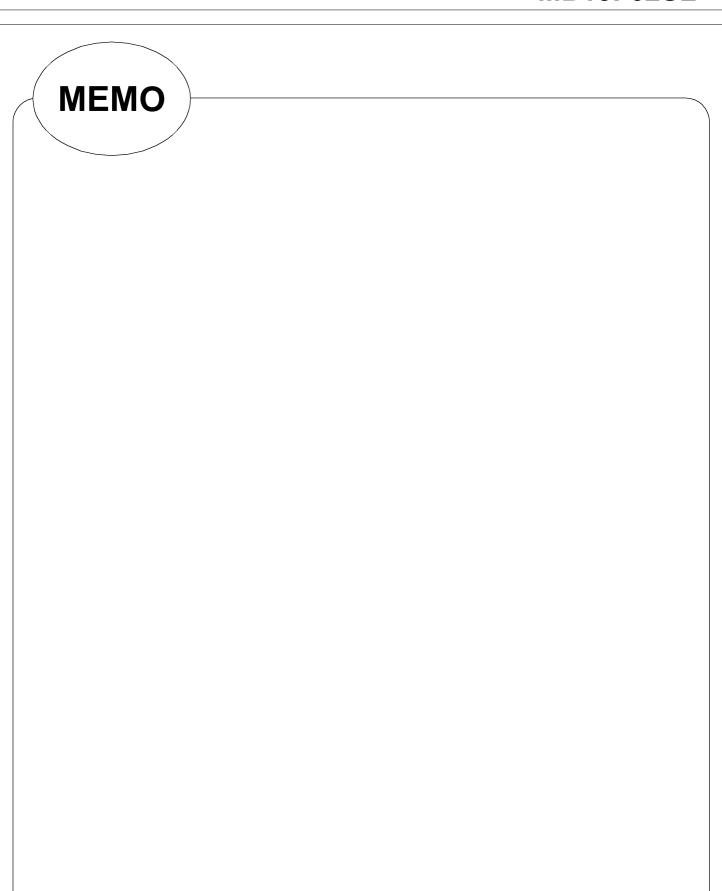


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