

LMC7101

Tiny Low Power Operational Amplifier with Rail-To-Rail Input and Output

General Description

The LMC7101 is a high performance CMOS operational amplifier available in the space saving SOT 23-5 Tiny package. This makes the LMC7101 ideal for space and weight critical designs. The performance is similar to a single amplifier of the LMC6482/6484 type, with rail-to-rail input and output, high open loop gain, low distortion, and low supply currents.

The main benefits of the Tiny package are most apparent in small portable electronic devices, such as mobile phones, pagers, notebook computers, personal digital assistants, and PCMCIA cards. The tiny amplifiers can be placed on a board where they are needed, simplifying board layout.

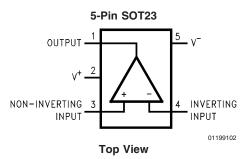
Features

- Tiny SOT23-5 package saves space—typical circuit layouts take half the space of SO-8 designs
- Guaranteed specs at 2.7V, 3V, 5V, 15V supplies
- Typical supply current 0.5 mA at 5V
- Typical total harmonic distortion of 0.01% at 5V
- 1.0 MHz gain-bandwidth
- Similar to popular LMC6482/6484
- Rail-to-rail input and output

Applications

- Mobile communications
- Notebooks and PDAs
- Battery powered products
- Sensor interface

Connection Diagram



Ordering Information

Package	Part Number	Package Marking	Transport Media	NSC Drawing	
	LMC7101AIM5	A00A	1k Units on Tape and Reel		Ī
5-Pin SOT23	LMC7101AIM5X	A00A	3k Units Tape and Reel	MF05A	
5-7111 30123	LMC7101BIM5	A00B	1k Units on Tape and Reel	IVIFUSA	
	LMC7101BIM5X	A00B	3k Units Tape and Reel		

Absolute Maximum Ratings (Note 1)

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/ Distributors for availability and specifications.

ESD Tolerance (Note 2) 2000V

Difference Input Voltage \pm Supply Voltage

Voltage at Input/Output Pin $(V^+) + 0.3V$, $(V^-) - 0.3V$ Supply Voltage $(V^+ - V^-)$ 16V

Current at Input Pin ± 5 mA

Current at Output Pin (Note 3) ± 35 mA

Current at Power Supply Pin 35 mA

Lead Temp. (Soldering, 10 sec.) 260°C

Storage Temperature Range -65°C to +150°C Junction Temperature (Note 4) 150°C

Recommended Operating Conditions (Note 1)

Supply Voltage $2.7V \le V^+ \le 15.5V$

Temperature Range

LMC7101AI, LMC7101BI -40°C to +85°C

Thermal Resistance (θ_{IA})

5-Pin STO23 325°C/W

2.7V Electrical Characteristics

Unless otherwise specified, all limits guaranteed for $T_J = 25^{\circ}C$, $V^+ = 2.7V$, $V^- = 0V$, $V_{CM} = V_O = V^+/2$ and $R_L > 1$ M Ω . **Boldface** limits apply at the temperature extremes.

			Тур	LMC7101AI	LMC7101BI	
Symbol	Parameter	Conditions	(Note 5)	Limit	Limit	Units
				(Note 6)	(Note 6)	
V _{os}	Input Offset Voltage	V ⁺ = 2.7V	0.11	6	9	mV
						max
TCV _{OS}	Input Offset Voltage		1			μV/°C
	Average Drift					
I _B	Input Bias Current		1.0	64	64	pA max
I _{os}	Input Offset Current		0.5	32	32	pA max
R _{IN}	Input Resistance		>1			Tera Ω
CMRR	Common-Mode	$0V \le V_{CM} \le 2.7V$	70	55	50	dB
	Rejection Ratio	V ⁺ = 2.7V				min
V _{CM}	Input Common-Mode Voltage	For CMRR ≥ 50 dB	0.0	0.0	0.0	V
	Range		0.0	0.0	0.0	min
			3.0	2.7	2.7	V
						max
PSRR	Power Supply	$V^+ = 1.35V$ to 1.65V				dB
	Rejection Ratio	$V^- = -1.35V$ to $-1.65V$	60	50	45	min
		V _{CM} = 0				
C_{IN}	Common-Mode Input		3			pF
	Capacitance					
Vo	Output Swing	$R_L = 2 k\Omega$	2.45	2.15	2.15	V min
			0.25	0.5	0.5	V max
		$R_L = 10 \text{ k}\Omega$	2.68	2.64	2.64	V min
			0.025	0.06	0.06	V max
I _S	Supply Current		0.5	0.81	0.81	mA
				0.95	0.95	max
SR	Slew Rate (Note 8)		0.7			V/µs
GBW	Gain-Bandwidth Product		0.6			MHz

3V DC Electrical Characteristics

Unless otherwise specified, all limits guaranteed for $T_J=25^{\circ}C$, $V^+=3V$, $V^-=0V$, $V_{CM}=1.5V$, $V_O=V^+/2$ and $R_L=1~M\Omega$. **Boldface** limits apply at the temperature extremes.

			Тур	LMC7101AI	LMC7101BI	
Symbol	Parameter	Conditions	(Note 5)	Limit	Limit	Units
				(Note 6)	(Note 6)	
V _{OS}	Input Offset Voltage		0.11	4	7	mV
				6	9	max
TCV _{OS}	Input Offset Voltage Average Drift		1			μV/°C
I _B	Input Current		1.0	64	64	pA max
I _{os}	Input Offset Current		0.5	32	32	pA max
R _{IN}	Input Resistance		>1			Tera Ω
CMRR	Common-Mode Rejection	$0V \le V_{CM} \le 3V$	74	64	60	db
	Ratio	V ⁺ = 3V				min
V _{CM}	Input Common-Mode Voltage	For CMRR ≥ 50 dB	0.0	0.0	0.0	V
	Range					min
			3.3	3.0	3.0	V
						max
PSRR	Power Supply Rejection Ratio	$V^+ = 1.5V$ to 7.5V	80	68	60	dB
		$V^- = -1.5V$ to $-7.5V$				min
		$V_O = V_{CM} = 0$				
C _{IN}	Common-Mode Input Capacitance		3			pF
Vo	Output Swing	$R_L = 2 k\Omega$	2.8	2.6	2.6	V min
			0.2	0.4	0.4	V max
		$R_L = 600\Omega$	2.7	2.5	2.5	V min
			0.37	0.6	0.6	V max
Is	Supply Current		0.5	0.81	0.81	mA
				0.95	0.95	max

5V DC Electrical Characteristics

Unless otherwise specified, all limits guaranteed for $T_J=25^{\circ}C$, $V^+=5V$, $V^-=0V$, $V_{CM}=1.5V$, $V_O=V^+/2$ and $R_L=1~M\Omega$. **Boldface** limits apply at the temperature extremes.

				Тур	LMC7101AI	LMC7101BI	
Symbol	Parameter	Condition	าร	(Note 5)	Limit	Limit	Units
-					(Note 6)	(Note 6)	
V _{os}	Input Offset Voltage	V ⁺ = 5V		0.11	3	7	mV
					5	9	max
TCV _{os}	Input Offset Voltage			1.0			μV/°C
	Average Drift						
I _B	Input Current			1	64	64	pA max
I _{os}	Input Offset Current			0.5	32	32	pA max
R _{IN}	Input Resistance			>1			Tera Ω
CMRR	Common-Mode	$0V \le V_{CM} \le 5V$		82	65	60	db
	Rejection Ratio				60	55	min
+PSRR	Positive Power Supply	V ⁺ = 5V to 15V		82	70	65	dB
	Rejection Ratio	$V^- = 0V, V_O = 1.5V$			65	62	min
-PSRR	Negative Power Supply	$V^{-} = -5V \text{ to } -15V$		82	70	65	dB
	Rejection Ratio	$V^{+} = 0V, V_{O} = -1.5V$	•		65	62	min
V _{CM}	Input Common-Mode	For CMRR ≥ 50 dB		-0.3	-0.20	-0.20	V
	Voltage Range				0.00	0.00	min
				5.3	5.20	5.20	V
					5.00	5.00	max
C _{IN}	Common-Mode			3			pF
	Input Capacitance						
Vo	Output Swing	$R_L = 2 k\Omega$		4.9	4.7	4.7	V
					4.6	4.6	min
				0.1	0.18	0.18	V
					0.24	0.24	max
		$R_L = 600\Omega$		4.7	4.5	4.5	V
					4.24	4.24	min
				0.3	0.5	0.5	V
					0.65	0.65	max
I _{sc}	Output Short Circuit	$V_O = 0V$	Sourcing	24	16	16	mA
	Current				11	11	min
		$V_O = 5V$	Sinking	19	11	11	mA
					7.5	7.5	min
l _s	Supply Current			0.5	0.85	0.85	mA
					1.0	1.0	max

5V AC Electrical Characteristics

Unless otherwise specified, all limits guaranteed for $T_J = 25^{\circ}C$, $V^+ = 5V$, $V^- = 0V$, $V_{CM} = 1.5V$, $V_O = V^+/2$ and $R_L = 1~M\Omega$. **Boldface** limits apply at the temperature extremes.

			Тур	LMC7101AI	LMC7101BI	
Symbol	Parameter	Conditions	(Note 5)	Limit	Limit	Units
				(Note 6)	(Note 6)	
THD	Total Harmonic	$F = 10 \text{ kHz}, A_V = -2$	0.01			%
	Distortion	$R_L = 10 \text{ k}\Omega, V_O = 4.0 V_{PP}$				
SR	Slew Rate		1.0			V/µs
GBW	Gain Bandwidth Product		1.0			MHz

15V DC Electrical Characteristics

Unless otherwise specified, all limits guaranteed for $T_J=25^{\circ}C$, $V^+=15V$, $V^-=0V$, $V_{CM}=1.5V$, $V_O=V^+/2$ and $R_L=1~M\Omega$. **Boldface** limits apply at the temperature extremes.

				Тур	LMC7101AI	LMC7101BI	
Symbol	Parameter	Condi	tions	(Note 5)	Limit	Limit	Units
					(Note 6)	(Note 6)	
Vos	Input Offset Voltage			0.11			mV max
TCVos	Input Offset Voltage			1.0			μV/°C
	Average Drift						
I _B	Input Current			1.0	64	64	pA max
I _{os}	Input Offset Current			0.5	32	32	pA max
R _{IN}	Input Resistance			>1			Tera Ω
CMRR	Common-Mode	0V ≤ V _{CM} ≤ 15V		82	70	65	dB
	Rejection Ratio				65	60	min
+PSRR	Positive Power Supply	V ⁺ = 5V to 15V		82	70	65	dB
	Rejection Ratio	$V^- = 0V, V_O = 1.8$	5V		65	62	min
-PSRR	Negative Power Supply	$V^- = -5V \text{ to } -15V$		82	70	65	dB
	Rejection Ratio	$V^{+} = 0V, V_{O} = -1$.5V		65	62	min
V _{CM}	Input Common-Mode	V ⁺ = 5V		-0.3	-0.20	-0.20	V
5.1.	Voltage Range	For CMRR ≥ 50 dB			0.00	0.00	min
				15.3	15.20	15.20	V
					15.00	15.00	max
A _V	Large Signal Voltage Gain	e Signal Voltage Gain $R_L = 2 kΩ$ Sourcing 3^2	340	80	80		
·	(Note 7)	_			40	40	
			Sinking	24	15	15	
					10	10	V/mV
		$R_L = 600\Omega$	Sourcing	300	34	34	
			Sinking	15	6	6	
C_{IN}	Input Capacitance			3			pF
Vo	Output Swing	V ⁺ = 15V		14.7	14.4	14.4	V
		$R_L = 2 k\Omega$			14.2	14.2	min
				0.16	0.32	0.32	V
					0.45	0.45	max
		V ⁺ = 15V		14.1	13.4	13.4	V
		$R_L = 600\Omega$			13.0	13.0	min
				0.5	1.0	1.0	V
					1.3	1.3	max
I _{sc}	Output Short Circuit Current	V _O = 0V	Sourcing	50	30	30	
	(Note 9)				20	20	mA
		V _O = 12V	Sinking	50	30	30	min
					20	20	
I _s	Supply Current			0.8	1.50	1.50	mA
					1.71	1.71	max

15V AC Electrical Characteristics

Unless otherwise specified, all limits guaranteed for $T_J = 25^{\circ}C$, $V^+ = 15V$, $V^- = 0V$, $V_{CM} = 1.5V$, $V_O = V^+/2$ and $R_L = 1~M\Omega$. **Boldface** limits apply at the temperature extremes.

			Тур	LMC7101AI	LMC7101BI	
Symbol	Parameter	Conditions	(Note 5)	Limit	Limit	Units
				(Note 6)	(Note 6)	
SR	Slew Rate	V ⁺ = 15V	1.1	0.5	0.5	V/µs
	(Note 8)			0.4	0.4	min
GBW	Gain-Bandwidth Product	V ⁺ = 15V	1.1			MHz
φ _m	Phase Margin		45			deg
G _m	Gain Margin		10			dB
e _n	Input-Referred Voltage Noise	f = 1 kHz, V _{CM} = 1V	37			$\frac{\text{nV}}{\sqrt{\text{Hz}}}$
i _n	Input-Referred Current Noise	f = 1 kHz	1.5			$\frac{fA}{\sqrt{Hz}}$
THD	Total Harmonic Distortion	f = 10 kHz, A _V = -2	0.01			%
		$R_L = 10 \text{ k}\Omega, V_O = 8.5 V_{PP}$				

Note 1: Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is intended to be functional, but specific performance is not guaranteed. For guaranteed specifications and the test conditions, see the Electrical Characteristics.

Note 2: Human Body Model is 1.5 k Ω in series with 100 pF.

Note 3: Applies to both single-supply and split-supply operation. Continuous short operation at elevated ambient temperature can result in exceeding the maximum allowed junction temperature at 150°C.

Note 4: The maximum power dissipation is a function of $T_{J(MAX)}$, θ_{JA} and T_A . The maximum allowable power dissipation at any ambient temperature is $P_D = (T_{J(MAX)} - T_A)/\theta_{JA}$. All numbers apply for packages soldered directly into a PC board.

Note 5: Typical Values represent the most likely parametric norm.

Note 6: All limits are guaranteed by testing or statistical analysis.

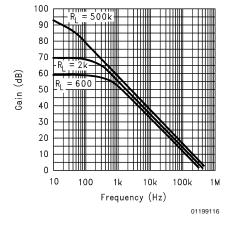
Note 7: $V^+ = 15V$, $V_{CM} = 1.5V$ and R_L connect to 7.5V. For sourcing tests, 7.5V $\leq V_O \leq 12.5V$. For sinking tests, 2.5V $\leq V_O \leq 7.5V$.

Note 8: V^+ = 15V. Connected as a voltage follower with a 10V step input. Number specified is the slower of the positive and negative slew rates. R_L = 100 k Ω connected to 7.5V. Amp excited with 1 kHz to produce V_Q = 10 V_{PP} .

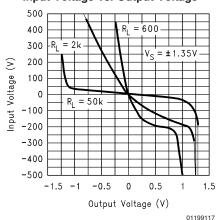
Note 9: Do not short circuit output to V⁺ when V⁺ is greater than 12V or reliability will be adversely affected.

2.7V Typical Performance Characteristics $V^+ = 2.7V$, $V^- = 0V$, $T_A = 25$ °C, unless otherwise specified.



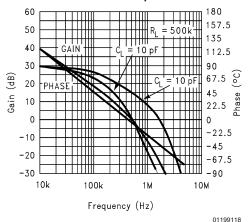


Input Voltage vs. Output Voltage

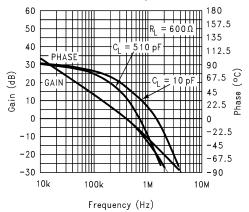


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Gain and Phase vs. Capacitance Load

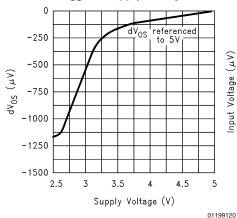


Gain and Phase vs. Capacitance Load

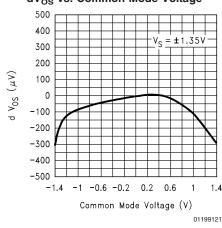


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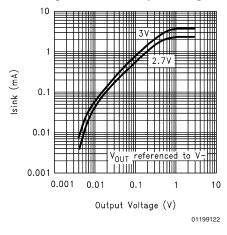
dV_{OS} vs. Supply Voltage



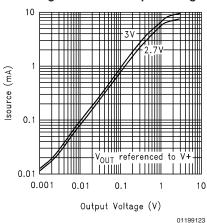
$\mathrm{dV}_{\mathrm{OS}}$ vs. Common Mode Voltage



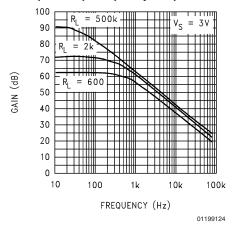
Sinking Current vs. Output Voltage



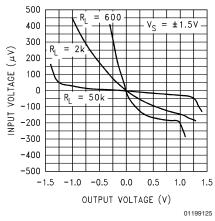
Sourcing Current vs. Output Voltage



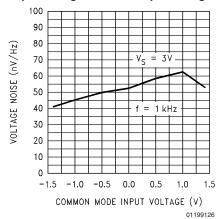
Open Loop Frequency Response



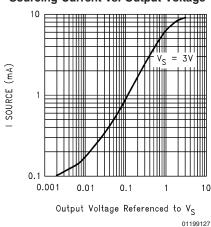
Input Voltage vs. Output Voltage



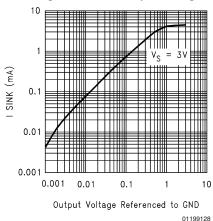
Input Voltage Noise vs. Input Voltage



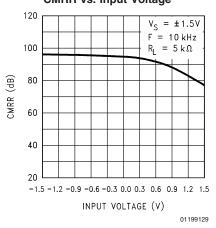
Sourcing Current vs. Output Voltage



Sinking Current vs. Output Voltage

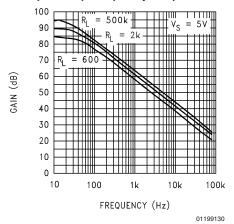


CMRR vs. Input Voltage

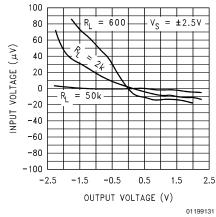


5V Typical Performance Characteristics $V^+ = 5V$, $V^- = 0V$, $T_A = 25$ °C, unless otherwise specified.

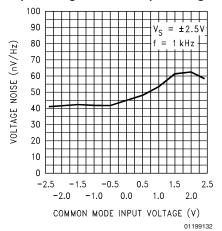
Open Loop Frequency Response



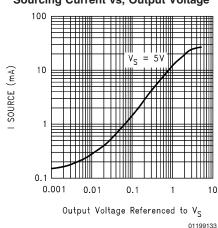
Input Voltage vs. Output Voltage



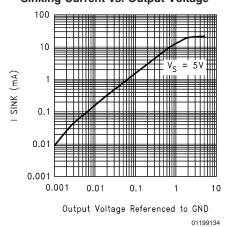
Input Voltage Noise vs. Input Voltage



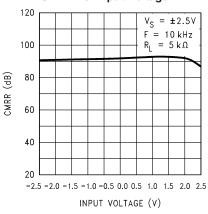
Sourcing Current vs, Output Voltage



Sinking Current vs. Output Voltage

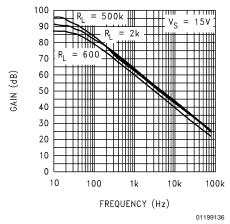


CMRR vs. Input Voltage

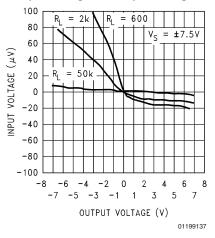


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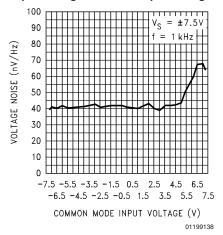
Open Loop Frequency Response



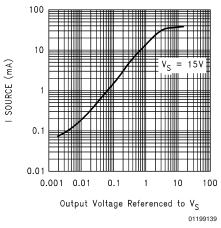
Input Voltage vs. Output Voltage



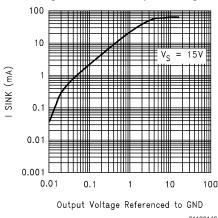
Input Voltage Noise vs. Input Voltage



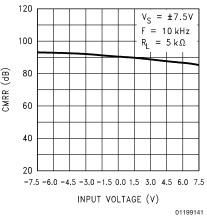
Sourcing Current vs. Output Voltage



Sinking Current vs. Output Voltage

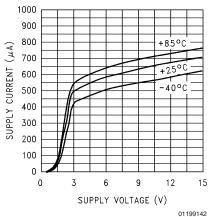


CMRR vs. Input Voltage

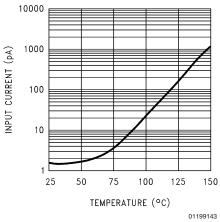


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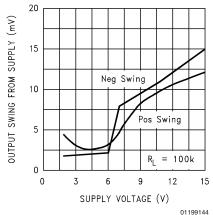
Supply Current vs. Supply Voltage



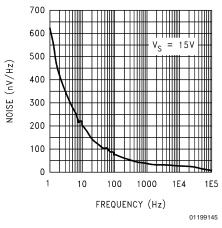
Input Current vs. Temperature



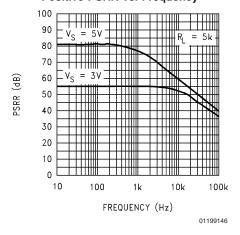
Output Voltage Swing vs. Supply Voltage



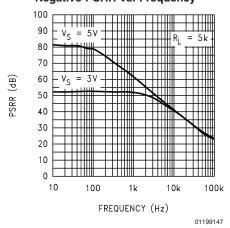
Input Voltage Noise vs. Frequency



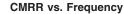
Positive PSRR vs. Frequency

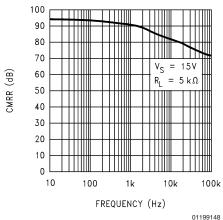


Negative PSRR vs. Frequency

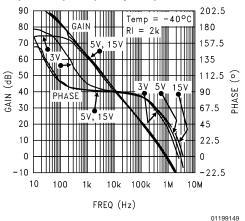


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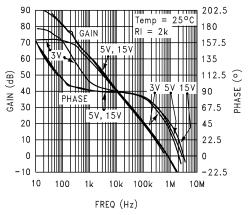




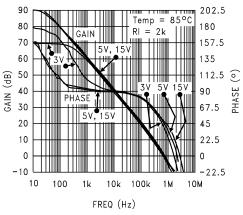
Open Loop Frequency Response @ -40°C



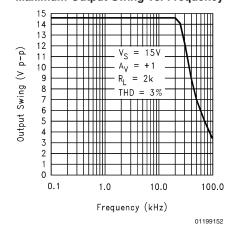
Open Loop Frequency Response @ 25°C



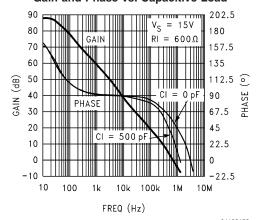
Open Loop Frequency Response @ 85°C



Maximum Output Swing vs. Frequency



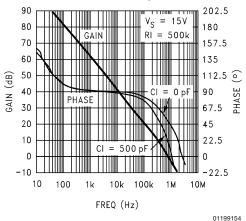
Gain and Phase vs. Capacitive Load

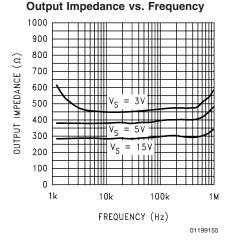


01199153

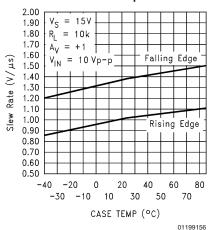
specified. (Continued)

Gain and Phase vs. Capacitive Load

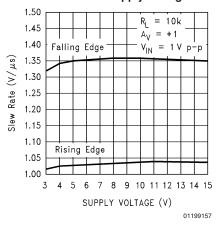




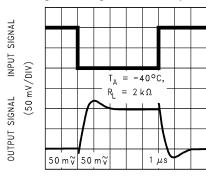
Slew Rate vs. Temperature



Slew Rate vs. Supply Voltage



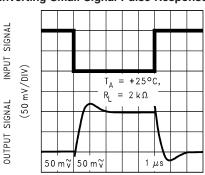
Inverting Small Signal Pulse Response



TIME (1 µs/DIV)

01199158

Inverting Small Signal Pulse Response

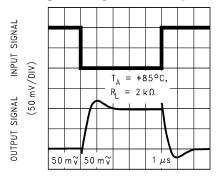


TIME (1 µs/DIV)

01199159

specified. (Continued)

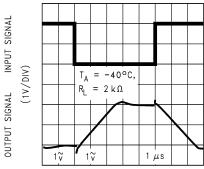
Inverting Small Signal Pulse Response



TIME $(1 \mu s/DIV)$

01199160

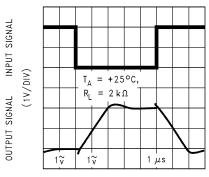
Inverting Large Signal Pulse Response



TIME $(1 \mu s/DIV)$

1100161

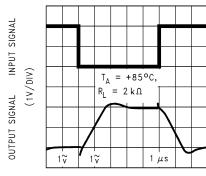
Inverting Large Signal Pulse Response



TIME $(1 \mu s/DIV)$

01199162

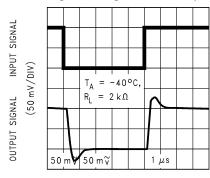
Inverting Large Signal Pulse Response



TIME $(1 \mu s/DIV)$

01199163

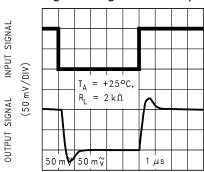
Non-Inverting Small Signal Pulse Response



TIME $(1 \mu s/DIV)$

01199164

Non-Inverting Small Signal Pulse Response

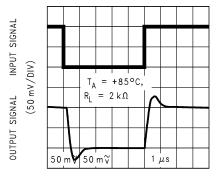


TIME $(1\mu s/DIV)$

01199165

specified. (Continued)

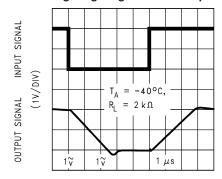
Non-Inverting Small Signal Pulse Response



TIME $(1 \mu s/DIV)$

01199166

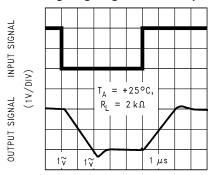
Non-Inverting Large Signal Pulse Response



TIME $(1 \mu s/DIV)$

1199167

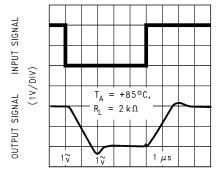
Non-Inverting Large Signal Pulse Response



TIME (1 μ s/DIV)

01199168

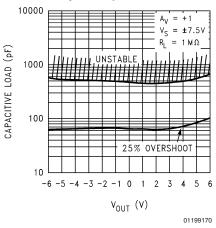
Non-Inverting Large Signal Pulse Response



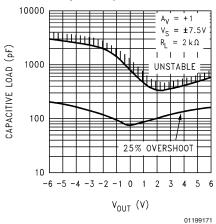
TIME (1 µs/DIV)

01199169

Stability vs. Capacitive Load

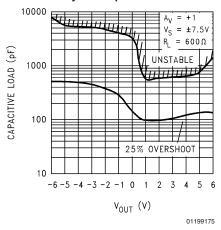


Stability vs. Capacitive Load

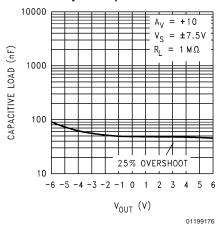


specified. (Continued)

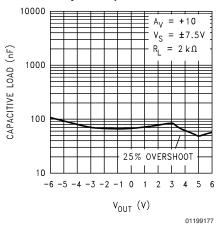
Stability vs. Capacitive Load



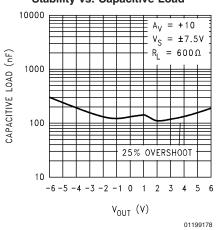
Stability vs. Capacitive Load



Stability vs. Capacitive Load



Stability vs. Capacitive Load



Application Information

1.0 BENEFITS OF THE LMC7101 TINY AMP

Size

The small footprint of the SOT 23-5 packaged Tiny amp, (0.120 x 0.118 inches, 3.05 x 3.00 mm) saves space on printed circuit boards, and enable the design of smaller electronic products. Because they are easier to carry, many customers prefer smaller and lighter products.

Height

The height (0.056 inches, 1.43 mm) of the Tiny amp makes it possible to use it in PCMCIA type III cards.

Signal Integrity

Signals can pick up noise between the signal source and the amplifier. By using a physically smaller amplifier package, the Tiny amp can be placed closer to the signal source, reducing noise pickup and increasing signal integrity. The Tiny amp can also be placed next to the signal destination, such as a buffer for the reference of an analog to digital converter.

Simplified Board Layout

The Tiny amp can simplify board layout in several ways. First, by placing an amp where amps are needed, instead of routing signals to a dual or quad device, long pc traces may be avoided.

By using multiple Tiny amps instead of duals or quads, complex signal routing and possibly crosstalk can be reduced.

Low THD

The high open loop gain of the LMC7101 amp allows it to achieve very low audio distortion—typically 0.01% at 10 kHz with a 10 k Ω load at 5V supplies. This makes the Tiny an excellent for audio, modems, and low frequency signal processing.

Low Supply Current

The typical 0.5 mA supply current of the LMC7101 extends battery life in portable applications, and may allow the reduction of the size of batteries in some applications.

Wide Voltage Range

The LMC7101 is characterized at 15V, 5V and 3V. Performance data is provided at these popular voltages. This wide voltage range makes the LMC7101 a good choice for devices where the voltage may vary over the life of the batteries.

2.0 INPUT COMMON MODE

Voltage Range

The LMC7101 does not exhibit phase inversion when an input voltage exceeds the negative supply voltage. *Figure 1* shows an input voltage exceeding both supplies with no resulting phase inversion of the output.

The absolute maximum input voltage is 300 mV beyond either rail at room temperature. Voltages greatly exceeding this maximum rating, as in *Figure 2*, can cause excessive current to flow in or out of the input pins, adversely affecting reliability.

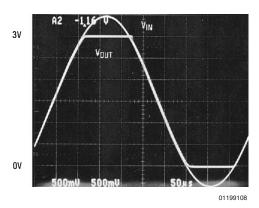


FIGURE 1. An Input Voltage Signal Exceeds the LMC7101 Power Supply Voltages with No Output Phase Inversion

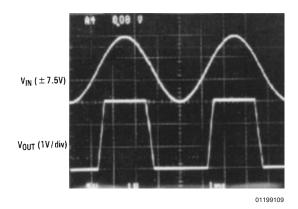


FIGURE 2. A ±7.5V Input Signal Greatly Exceeds the 3V Supply in *Figure 3* Causing No Phase Inversion Due to R_I

Applications that exceed this rating must externally limit the maximum input current to ± 5 mA with an input resistor as shown in *Figure 3*.

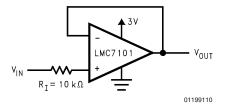


FIGURE 3. R_I Input Current Protection for Voltages Exceeding the Supply Voltage

3.0 RAIL-TO-RAIL OUTPUT

The approximate output resistance of the LMC7101 is 180Ω sourcing and 130Ω sinking at $V_S=3V$ and 110Ω sourcing and 80Ω sinking at $V_S=5V$. Using the calculated output resistance, maximum output voltage swing can be estimated as a function of load.

Application Information (Continued)

4.0 CAPACITIVE LOAD TOLERANCE

The LMC7101 can typically directly drive a 100 pF load with $V_{\rm S}=15V$ at unity gain without oscillating. The unity gain follower is the most sensitive configuration. Direct capacitive loading reduces the phase margin of op-amps. The combination of the op-amp's output impedance and the capacitive load induces phase lag. This results in either an underdamped pulse response or oscillation.

Capacitive load compensation can be accomplished using resistive isolation as shown in *Figure 4*. This simple technique is useful for isolating the capacitive input of multiplexers and A/D converters.

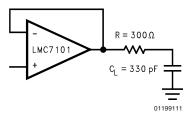


FIGURE 4. Resistive Isolation of a 330 pF Capacitive Load

5.0 COMPENSATING FOR INPUT CAPACITANCE WHEN USING LARGE VALUE FEEDBACK RESISTORS

When using very large value feedback resistors, (usually $> 500 \text{ k}\Omega$) the large feed back resistance can react with the input capacitance due to transducers, photodiodes, and circuit board parasitics to reduce phase margins.

The effect of input capacitance can be compensated for by adding a feedback capacitor. The feedback capacitor (as in *Figure 5*), C_f is first estimated by:

$$\frac{1}{2\pi R_1\,C_{IN}} \geq \frac{1}{2\pi R_2 C_f}$$

or $R_1 \ C_{IN} \leq R_2 \ C_f$

which typically provides significant overcompensation.

Printed circuit board stray capacitance may be larger or smaller than that of a breadboard, so the actual optimum value for $C_{\rm F}$ may be different. The values of $C_{\rm F}$ should be checked on the actual circuit. (Refer to the LMC660 quad CMOS amplifier data sheet for a more detailed discussion.)

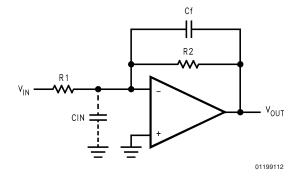


FIGURE 5. Cancelling the Effect of Input Capacitance

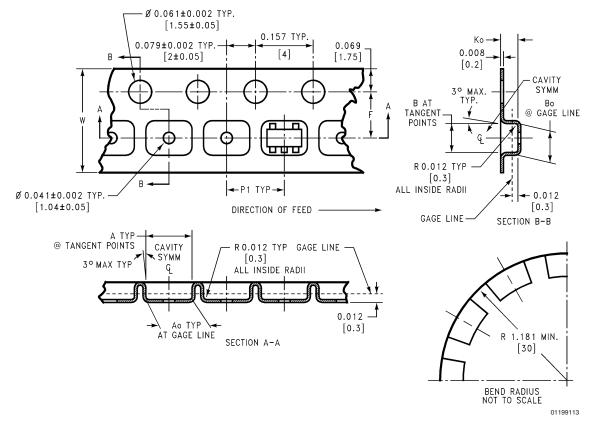
Application Information (Continued)

SOT-23-5 TAPE AND REEL SPECIFICATION

Tape Format

Tape Section	Tape Section # Cavities		Cover Tape Status	
Leader	0 (min)	Empty	Sealed	
(Start End)	art End) 75 (min) Empty		Sealed	
Carrier	3000	Filled	Sealed	
	1000	Filled	Sealed	
Trailer	125 (min)	Empty	Sealed	
(Hub End)	0 (min)	Empty	Sealed	

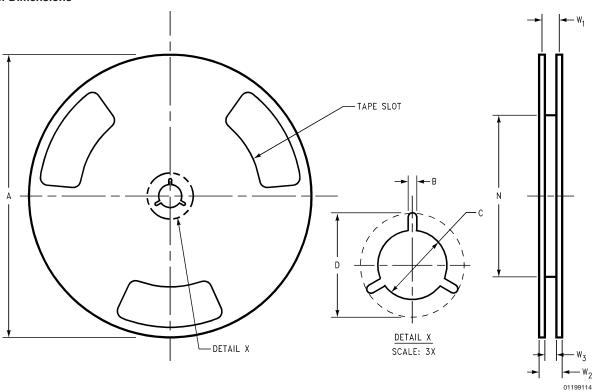
Tape Dimensions



8 mm	0.130	0.124	0.130	0.126	0.138 ±0.002	0.055 ±0.004	0.157	0.315 ±0.012
	(3.3)	(3.15)	(3.3)	(3.2)	(3.5 ±0.05)	(1.4 ±0.11)	(4)	(8 ±0.3)
Гаре Size	DIM A	DIM Ao	DIM B	DIM Bo	DIM F	DIM Ko	DIM P1	DIM W

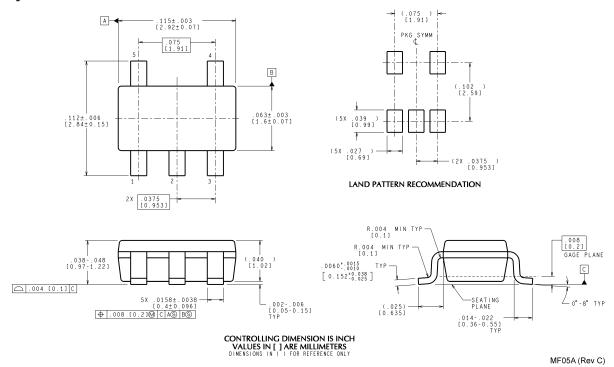
Application Information (Continued)

Reel Dimensions



8 mm	7.00	0.059	0.512	0.795	2.165	0.331 + 0.059/-0.000	0.567	W1+ 0.078/-0.039
	330.00	1.50	13.00	20.20	55.00	8.40 + 1.50/-0.00	14.40	W1 + 2.00/-1.00
Tape Size	Α	В	С	D	N	W1	W2	W3

Physical Dimensions inches (millimeters) unless otherwise noted



5-Pin SOT23 Package **NS Package Number MF05A**

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